

DC MOTOR WITH BRUSHES

[0000]

This application claims priority to Japanese patent application serial number 2002-378643, the contents of which are incorporated herein by reference.

[0001]

BACKGROUND OF THE INVENTION

Technical Field

The present invention relates to DC (direct current) motors having brushes for supplying power to the DC motors.

[0002]

Description of the Related Art

Up to now, DC motors having brushes have been used for various machines and apparatus, e.g., vehicles, because the DC motors have a relatively high efficiency and can be easily controlled.

[0003]

FIG. 4(A) schematically illustrates a known DC motor that has been formed by using a concentrated winding method. The known DC motor is configured as a four-pole and six-slot type of DC motor that has four permanent magnets and six slots, i.e., six coils (delta connection type) and six cores around which the coils are wound. Each of the permanent magnets has an N-pole surface and an S-pole surface. The terminals of each coil are connected to two corresponding commutator segments. In general, the number of coils is set to be larger than the number of poles (permanent magnets) in order to help minimize dead points during rotation of the armature.

[0004]

More specifically, in the known DC motor shown in FIG. 4(A), permanent magnets M1 - M4 are fixed to an inner wall of a yoke in order to constitute a stator. The quantity of individual permanent magnets is chosen to be even in number. For example, in the known DC motor four poles (permanent magnets) M1 to M4 are arranged in a circle to substantially form a cylinder. The polarity of the surface of each of the poles M1 - M4, on the side facing the rotor, alternates from magnet to adjoining magnet. Thus, the polarity of a surface directly opposing

the rotor of the magnet M1 is N, and the polarity of same surfaces of the permanent magnets M2 and M4, both of which adjoin permanent magnet M1, is S.

[0005]

The rotor includes cores T1 - T6 with slots R1 - R6 formed therebetween. Coils C1 - C6 are wound around the corresponding cores T1 - T6 via the corresponding slots. Commutator segments S1 - S6 are connected to the terminals of the corresponding coils. The commutator segments S1 - S6 are electrically insulated from each other. The rotor rotates about the center axis that defines the stator. The rotor rotates within the stator.

[0006]

Brushes B1 - B4 are disposed and are arranged such that the individual commutator segments S1 - S6 form an electrical connection with various individual brushes B1 - B4 when the commutator segments S1 - S6 rotate through a predetermined angle. The number of the brushes B1 - B4 is the same or smaller than the number of the poles (permanent magnets). In the configuration shown in FIG. 4(A), the number of the brushes (four) is equal to the number of the poles. A power source E is connected to each of the brushes B1 to B4 via a positive or negative terminal, so that the brushes supply current to the coils via the commutator segments that contact the brushes.

[0007]

FIG. 4(B) illustrates an example of positional relations between the brushes B1 - B4 of the DC motor shown in FIG. 4(A). In this example, the brushes B1 and B3 each extend approximately 20° around the center axis of the armature. Brush B1 begins from approximately a 25° position and extends in a counterclockwise direction (assuming the horizontal axis to the right of the rotor to be a 0° reference position). Brush B3 begins approximately 25° below a 180° reference position (205° from the 0° reference of B1) as shown in FIG. 4(B). Thus, each of the brushes B1 and B3 extends from an initial position indicated by DegB1s to a final position indicated by DegB1e. Similarly, the brushes B2 and B4 extend over a range of 20° from positions rotated 25° counterclockwise respectively from a 90° reference and 270° reference positions (105° and 295° from the 0° reference of B1) as shown in FIG. 4(B). Thus, each of the brushes B2 and B4 extends from an initial position indicated by DegB2s to a final position indicated by DegB2e.

[0008]

In the case of the DC motor shown in FIGS. 4(A) and 4(B), the rotor is rotated in the counterclockwise direction. The current is supplied to the coil C1 during the time when the boundary element between the commutator segments S1 and S2 is not within the 20° range corresponding to the contact area of any of the four brushes B1 - B4. Thus, during each rotation of the rotor, the coil C1 has four separate current supply periods. The supply periods correspond to when the S1 and S2 boundary element is rotated within the range established between DegB1e and DegB2s and the range established between DegB2e and DegB1s. The same general principle also applies to the other coils C2 - C6.

[0009]

Next, the magnetic flux density around each coil during one current supply period will be described with reference to FIGS. 5(A) and 5(B).

[0010]

FIG. 5(A) illustrates coil C1 having a boundary element positioned within an angular range between DegB1e (45° from the previous explanation), and the boundary between pole M1 and pole M2 (90°, as shown in FIG. 5(A)). In this state, a current is supplied to the coil C1 to produce a magnetic field (ϕ_c) within the corresponding core T1. The magnetic field (ϕ_c) thus produced extends in a direction away from the center of rotation of the rotor towards the pole M1. Simultaneously, a magnetic field (ϕ_m) is produced within the core T1 by the poles M1 and M2. The direction of the magnetic field (ϕ_m) produced by the poles M1 and M2 is inverse to the direction of the magnetic field (ϕ_c) produced by the electric current. The magnetic flux density generated by the current induced magnetic field (ϕ_c) around the coil C1 is reduced by the magnetic flux density of the magnetic field (ϕ_m) generated by the permanent magnets. Therefore, in the following descriptions, this angular range is called the "reduced density area."

[0011]

FIG. 5(B) illustrates the coil C1 positioned within an angular range from the change in polarity of the poles M1 to M2 (90° as shown in FIG. 5(B)), to a position where the supply of current for coil C1 is terminated at DegB2s (105° from the 0° reference). In this range, a current is supplied to the coil C1 to continue to produce a magnetic field (ϕ_c) within the corresponding core T1. The magnetic field (ϕ_c) thus produced, as explained previously,

[0023]

In a further aspect of the present teachings, each pole comprises a permanent magnet that has a first portion, defining the increased density area, and a second portion defining the reduced density area. The magnetic force of the first portion is smaller than the magnetic force of the second portion. For an example, the first and second portions may be made of different materials. Alternatively, the first and second portions may be magnetized to different magnetizing strengths.

[0024]

Also with this arrangement, many of the same effects regarding magnetic flux density can be realized by using a simple construction.

[0025]

In a still further aspect of the present teachings, each pole comprises a permanent magnet that has a first portion defining the increased density area and a second portion defining the reduced density area. A gap is defined between the permanent magnet and the end of the individual cores of the rotor. The gap defined by at least a part of the first portion is larger than the gap defined by the second portion. For example, forming a recess in the first portion may increase the gap.

[0026]

With this arrangement as in the previous aspects, use of a relatively simple construction produces many of the same effects regarding magnetic flux density.

[0027]

BRIEF DESCRIPTION OF THE DRAWINGS

Additional objects, features, and advantages, of the present invention will be readily understood after reading the following detailed description together with the claims and the accompanying drawings, in which:

FIGS. 1(A) and 1(B) are schematic views of a first and second representative DC motor; and

FIGS. 2(A), 2(B) and 2(C) are schematic views of a third representative DC motor; and

FIG. 3 is a graph illustrating a characteristic counter electromotive voltage (V) of a DC motor according to the present invention that is superimposed upon a dotted line representing

the characteristic counter electromotive voltage of the known DC motor, both in relation to the angle (°) of rotation of a rotor; and

FIGS. 4(A) and 4(B) are schematic views of the known DC motor; and

FIGS 5(A) and 5(B) are views illustrating a magnetic flux density around each coil when a current is supplied during one representative cyclic period; and

FIG. 6 is a graph illustrating a characteristic counter electromotive voltage (V) in relation to an angle (°) of rotation of a rotor for the known DC motor.

[0028]

DETAILED DESCRIPTION OF THE INVENTION

Each of the additional features and teachings disclosed above and below may be utilized separately or in conjunction with other features and teachings to provide improved DC motors, and improved methods of manufacturing and using such DC motors. Representative examples of the present invention, which utilize many of these additional features and teachings both separately and in conjunction with each other, will now be described in detail with reference to the attached drawings. This detailed description is merely intended to teach a person of skill in the art further details for practicing preferred aspects of the present teachings and is not intended to limit the scope of the invention. Only the claims define the scope of the claimed invention. Therefore, combinations of features and steps disclosed in the following detailed description may not be necessary to practice the invention in the broadest sense, and are instead taught merely to particularly describe representative examples of the invention. Moreover, various features of the representative examples and the dependent claims may be combined in ways that are not specifically enumerated in order to provide additional useful embodiments of the present teachings.

[0029]

First Representative Embodiment

A first representative embodiment will now be described with reference to FIGS. 1(A) and 1(B), which schematically show a first representative DC motor that is configured as a four-pole and six-slot DC brushed motor. The basic construction of the first representative embodiment is the same as the known DC motor shown in FIGS. 4(A) & 4(B) and 5(A) & 5(B). Therefore, elements that are similar to or identical with the known DC motor are labeled with the same reference numerals and an explanation of these elements may not be repeated.

[0038]

As shown FIG. 3, the magnetic flux density produced around the coils by the magnetic forces of the poles when the supply of current to the coils is interrupted (within the increased density areas) can be reduced by the first representative DC motor using a relatively simple construction. Therefore, the change of the counter electromotive voltage per unit angle of rotation of the rotor ($\Delta V_n / \Delta \phi_n$) can be reduced or minimized in comparison with the change in the counter electromotive voltage of the known DC motor ($\Delta V_z / \Delta \phi_z$) (shown by the dotted line in FIG. 3). As a result, the counter electromotive voltage that is produced when the supply of current to the coils is interrupted can be reduced or minimized so that potential discharge between the brushes and the commutator segments can subsequently be reduced or minimized. The result is that excessive wear of the brushes can be reduced.

[0039]

The counter electromotive voltage produced at the coils as described above is a voltage resultant from the combination of the counter electromotive voltage due to the change in the magnetic forces of the poles caused as the rotor rotates, and the counter electromotive voltage due to a change in the current flowing across the coils. Therefore, the counter electromotive voltage may change in proportion to change in over time of the magnetic flux that mainly flows through the cores.

[0040]

In the known DC motors, the magnetic flux that flows through the cores exceeds the saturated magnetic flux density or exceeds a magnetic flux density nearly equal to the saturated density. Therefore, the change per unit time of the magnetic flux density is small and the counter electromotive voltage produced due to changes in the magnetic forces of the poles is also very small. As a result, a large current may flow across the coils before interruption of the supply of current to the coils.

[0041]

Next, with respect to the change of the counter electromotive voltage caused when the supply of current is interrupted, the change of the magnetic forces of the poles is very small because the interruption of the supply of current occurs over a very short time. Therefore, the counter electromotive voltage is produced mainly by the reduction of the coil current, also occurring over a very short time. In the known DC motors, a large current flows across the

coils before the interruption of supply of current. Therefore, the counter electromotive voltage may have a relatively large value in order to cause such a large change in the coil current.

[0042]

In contrast, according to the first representative embodiment, the magnetic forces of the poles in the increased density areas are set to be of low strength. Therefore, before the interruption of the current, the magnetic flux is allowed to change by a relatively large amount to increase the counter electromotive voltage, so that the current flowing across the coils may be reduced.

[0043]

In addition, because the current flowing across the coils is low before the interruption of the supply of current, it is possible to also reduce the magnitude change of the counter electromotive voltage that may be produced by the reduction of the current when the current is interrupted. Therefore, potential discharges can be effectively reduced or minimized, so that the resistance of brushes against excessive wear can be improved.

[0044]

Second Representative Embodiment

A second representative embodiment is a modification of the first representative embodiment, in which each pole includes at least two permanent magnets. A second representative DC motor differs from the first representative DC motor in that the magnetic force in the increased density area of each pole is set to be of relatively low strength. Therefore, the second representative embodiment will be described with reference to the same drawings as in the first representative embodiment, in particular FIG. 1(A). Because the poles M1 to M4 of the second representative DC motor are essentially identical, only the pole M4 will be explained as a representative example and an explanation of the poles M1 to M3 will be omitted. Further, the explanation of the pole M4 of the second representative DC motor will be made only for the features different from the representative pole M4 of the first representative DC motor.

[0045]

In the second representative DC motor, the pole M4 is formed by the permanent magnets M4a and M4b in the same manner as the first representative DC motor. However, the

magnetic force of the permanent magnet M4b is less than the magnetic force of the permanent magnet M4a.

[0046]

This may be realized by magnetizing a magnetic material such that the magnetization of a portion corresponding to the increased density area is weaker than the magnetization of a portion corresponding to the reduced density area. Otherwise, different materials having different magnetic qualities can be respectively used for a part corresponding to the increased density area and a part corresponding to the reduced density area.

[0047]

Further, in order to reduce the magnetic flux density when the supply of current to the coils is interrupted, the permanent magnet M4b, having a smaller magnetic force than permanent magnet M4a, preferably extends by a slight distance counterclockwise beyond the boundary position (DegB2s) of the increased density area, in the rotational direction of the rotor. Alternatively, the permanent magnet M4b may also extend within at least a part of the increased density area, for example extending from a position just short of the current interrupting position to the current interrupting position.

[0048]

Further, in addition to the reduction of the magnetic flux density produced around the coils within the increased density areas, the magnetic flux density produced around the coils within the reduced density areas may be increased. In such a case, each pole is configured to produce a large magnetic force within at least a part of the corresponding reduced density area. In order to provide such a large magnetic force, the magnetization may be increased or the material selected for a portion of the pole may be chosen to have a relatively large magnetic force.

[0049]

Also with this second representative embodiment, the magnetic flux density produced around the coils by the magnet forces of the poles when the supply of current to the coils is interrupted (within the increased density areas) can be reduced. The resultant change of the counter electromotive voltage per unit angle of rotation of the rotor ($\Delta V_n / \Delta \phi_n$) can be reduced or minimized in comparison to the change in the counter electromotive voltage of the known DC motor ($\Delta V_z / \Delta \phi_z$) (shown as the dotted line in FIG. 3). As a result, the counter

electromotive voltage that is produced when the supply of current to the coils is interrupted can be reduced or minimized, so that the potential discharge between the brushes and the commutator segments can be reduced or minimized and excessive wear of the brushes can be reduced.

[0050]

Third Representative Embodiment

A third representative embodiment is another modification of the first representative embodiment. Thus, a third representative DC motor differs from the first representative DC motor in that the magnetic flux density produced around the coils is reduced by increasing a gap (L1) between each core and at least a part of the corresponding increased density area (i.e., the corresponding pole) as shown in FIGS. 2(A). The third representative DC motor can reduce the magnetic flux density produced around the coils without requiring the changing of the magnitude magnetic forces of the increased density areas. Also, the third representative DC motor can be made with only a single magnet for each of the poles M1 to M4. Because the poles M1 to M4 of the third representative DC motor are identical, only the representative pole M4 will be explained in detail and an explanation of the poles M1 to M3 may be omitted. Further, the explanation of the pole M4 of the third representative DC motor will be made for only those features which are different from the representative pole M4 of the first representative DC motor.

[0051]

Referring to FIG. 2(A), the pole M4 is formed by a single permanent magnet and is configured such that the gap (L1) between the pole M4 and the core T1, within the corresponding increased density area, increases when the supply of current to the corresponding coil is interrupted.

[0052]

As shown in FIG. 2(A), a part of the pole M4, around the DegB2s position where the supply of current to the coil is interrupted, is configured as a gently curved recess. With this configuration, the magnetic flux density produced around the coil can be reduced at the point when the supply of current to the coil is interrupted.

[0053]

In an alternative embodiment shown in FIG. 2(B), a part of the pole M4, around the DegB2s position where the supply of current to the coil is interrupted, is configured as a stepped recess. In another alternative embodiment shown in FIG. 2(C), a part of the pole M4, around the DegB2s position where the supply of current to the coil is interrupted, is configured as a substantially triangular recess. Such a portion of the pole M4 may be configured as any kind of geometric recess or combination of geometric recesses other than those shown in FIGS. 2(A) to 2(C).

[0054]

Preferably, the recess extends by a small angle beyond the boundary position (DegB2s position) of the increased density area in the counterclockwise rotational direction, e.g. the motor rotational direction, in order to reduce the magnetic flux density when the supply of current to the coil is interrupted.

[0055]

In the third representative embodiment shown in FIG. 2(A) and the alternative embodiments shown in FIGS. 2(B) and 2(C), the recess of the pole M4 extends from the starting position of the increased density area (i.e., the 90° position defining the boundary between the poles M1 and M4) to a position just beyond the DegB2s position by a small angle in the counterclockwise direction, where the supply of current to the coil is interrupted. However, the recess may extend within at least a part of the increased density area. Thus the recess may extend over a range from a position short of the DegB2s position to the DegB2s position.

[0056]

In addition, the third representative embodiment shown in FIG. 2(A) and the alternative embodiments shown in FIGS. 2(B) and 2(C) may also be configured to increase the magnetic flux density that is produced around the coils within the reduced density areas in addition to reducing the magnetic flux density produced around the coils within the increased density areas. For example, this may be realized by providing a projection on at least a part of each pole within the reduced density area in order to reduce the gap (L1) between the pole and the cores.

[0057]

Also with the third representative embodiment and the third representative alternative embodiments, the magnetic flux density produced around the coils by the magnetic forces of

of the brushes may be smaller than the number of the poles. For example, if the number of the coils is an even number, the number of the brushes may be reduced by short-circuiting between two commutator segments that are displaced from each other by an angle of 180° .